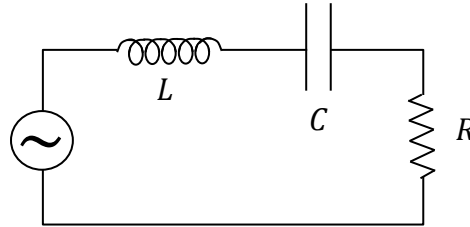
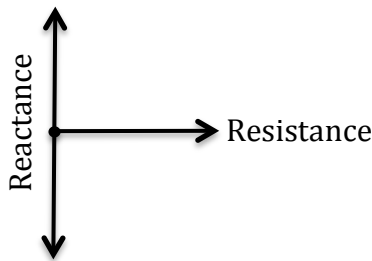


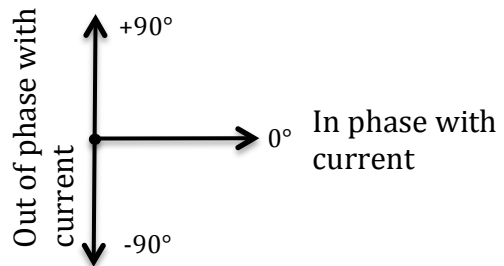
**Series RLC Filters and Geometry
or
The Mighty Impedance Diagram!**



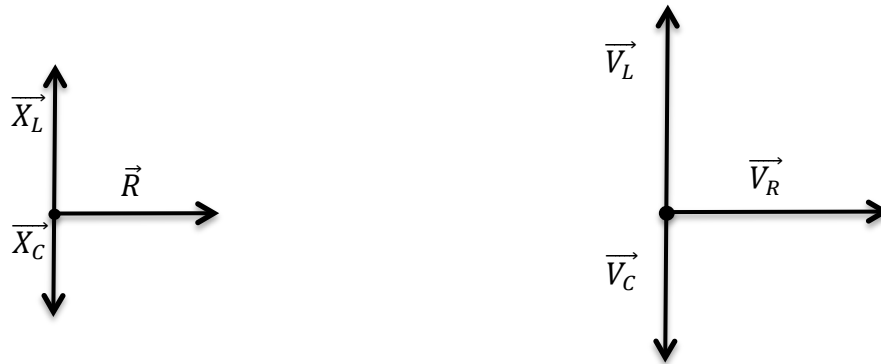
Impedances



Voltages



The impedance diagram on the left and the voltage diagram on the right will always be geometrically *similar*, meaning that all angles will be the same, and the relative lengths of vectors will be the same. In the following, we will draw the voltage diagram at a scale larger than that of the impedance diagram, just to emphasize this “similarity” property.



Where, for a given frequency f , the lengths of the reactance vectors are $X_L = 2\pi fL$ and $X_C = 1/(2\pi fC)$. (Note that we have drawn these diagrams assuming the inductor has a larger contribution than the capacitor, but the lengths may be different in general.)

Because of similarity, relative lengths are the same. So for example,

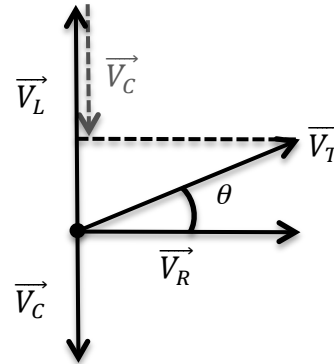
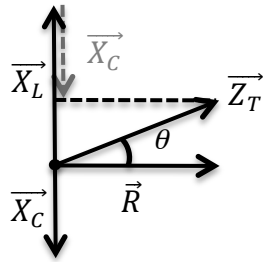
$$\frac{R}{X_L} = \frac{V_R}{V_L}$$

If we want to include the total impedance Z_T and total voltage V_T , these are simply the vector sums of their corresponding parts, i.e.

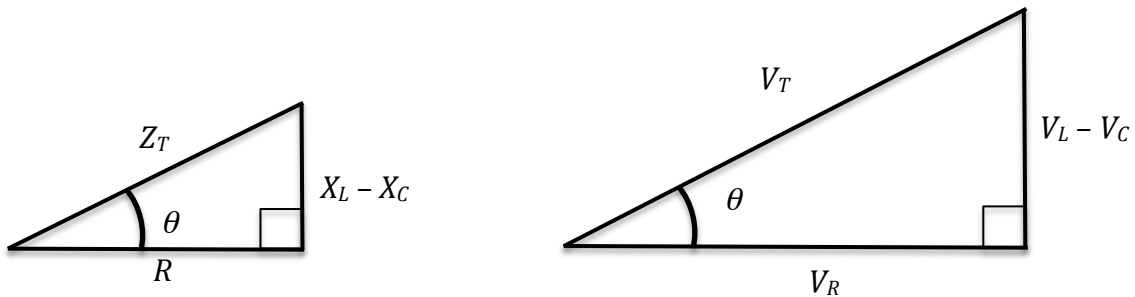
$$\vec{Z}_T = \vec{R} + \vec{X}_C + \vec{X}_L$$

$$\vec{V}_T = \vec{V}_R + \vec{V}_C + \vec{V}_L$$

as shown in the diagrams:



Where similarity tells us that θ is the same in both diagrams. Let us reduce the above diagrams into mere triangles, which we will “zoom in” on:



where vertical bars denote absolute value. (Note that we’ve drawn the triangles “upward,” but they could just as easily be drawn “downward” for $X_C > X_L$.) Using these triangles, let’s write out a few trigonometric identities:

$$Z_T = \sqrt{R^2 + (X_L - X_C)^2}$$

$$V_T = \sqrt{V_R^2 + (V_L - V_C)^2}$$

$\cos \theta$	$=$	$\frac{R}{Z_T}$	$=$	$\frac{V_R}{V_T}$ *
$\sin \theta$	$=$	$\frac{X_L - X_C}{Z_T}$	$=$	$\frac{V_L - V_C}{V_T}$
$\tan \theta$	$=$	$\frac{X_L - X_C}{R}$	$=$	$\frac{V_L - V_C}{V_R}$

The angle θ will be positive or negative depending on whether the total voltage is *leading* (+) or *lagging* (-) the current.

* Note that this is just a “voltage divider” formula, with R_T replaced by Z_T .

Example: Bandpass Filter

Say we measure the voltage across the resistor, i.e. we attach our “output” in series with L and C . The voltage gain G (or “transfer function”) of a filter is given by $G = V_{out}/V_{in}$, and in this case $V_{out} = V_R$ and $V_{in} = V_T$. So our G is simply

$$G = \frac{V_R}{V_T}$$

Using the impedance side of the diagrams above: $V_R/V_T = R/Z_T$. Thus G is

$$G = \frac{R}{\sqrt{R^2 + (X_L - X_C)^2}}$$

Let’s pretty this up: Multiply both the denominator and the numerator by $1/R$ to find

$$G = \left[1 + \left(\frac{X_C}{R} - \frac{X_L}{R} \right)^2 \right]^{-1/2} = \left[1 + \left(\frac{1}{2\pi f C R} - \frac{2\pi f L}{R} \right)^2 \right]^{-1/2}$$

Then we can group “variables that are not f ” into two new variables f_1 and f_2 such that

$$f_1 = \frac{1}{2\pi R C}$$

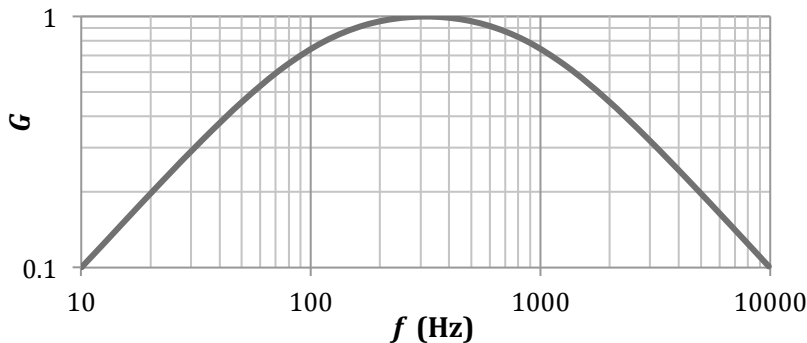
and

$$f_2 = \frac{R}{2\pi L}$$

With these substitutions, we arrive at

$$G = \left[1 + \left(\frac{f_1}{f} - \frac{f}{f_2} \right)^2 \right]^{-1/2} \tag{1}$$

To see this in action, let’s choose some values: $R = 8 \Omega$, $f_1 = 100 \text{ Hz}$ and $f_2 = 1000 \text{ Hz}$. Our choice of f_1 means that $C = 199 \mu\text{F}$ and our choice of f_2 means that $L = 1.27 \text{ mH}$. (Check: derive these values yourself!) The graph below shows a Bode (log-log) plot of the gain vs. frequency[†]:



Note that if $f_1 \rightarrow 0$ ($C \rightarrow \infty$), we have purely a high pass filter, and if $f_2 \rightarrow \infty$ ($L \rightarrow 0$) we have purely a low pass filter. Thus Eq. (1) includes low-, high-, and band-pass filters.

This result is only *one example* of the many things *you* can figure out using an impedance diagram and trigonometry!

Exercise: For the values in the above example, find f for which the total impedance is purely resistive, i.e., $Z_T = R$. What is the gain at this frequency? Answers below[†].

[†] What is often of interest may not be the voltage *per se* but rather the *power*, the gain for which will be the *square* of G . Thus the slope of the “sides” for the power graph will be ± 2 , not ± 1 as above.

[†]Answers: 316 Hz, G=1